Evaluation of Starting Current of Induction Motors Using Artificial Neural Network

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Induction motors (IMs) are widely used in industry including it be an electrical or not. However during starting period, their starting currents are so large that can damage equipment. Therefore, this current should be estimated accurately to prevent hazards caused by it. In this paper, the artificial neural network (ANN) as an intelligent tool is used to evaluate starting current peak of IMs. Both Multilayer Perceptron (MLP) and Radial Basis Function (RBF) structures have been analyzed. Six learning algorithms, backpropagation (BP), delta-bar-delta (DBD), extended delta-bar-delta (EDBD), directed random search (DRS), quick propagation (QP), and levenberg marquardt (LM) were used to train the MLP. The simulation results using MATLAB show that most developed ANNs can estimate the starting current peak of IMs with good accuracy. However, it is proven that LM and EDBD algorithms present better performance for starting current evaluation based on average of relative and absolute errors.

Index Terms: Induction motors, multilayer perceptron, radial basis function, starting current.

Nomenclature

$V_{\ ds}$, $V_{\ qs}$	direct/quadrature component of stator voltage
V $_{dr}^{\prime}$, V $_{qr}^{\prime}$ *	direct/quadrature component of rotor voltage
$\mathbf{i}_{ds}, \mathbf{i}_{qs}$	direct/quadrature component of stator current
$\mathbf{i}_{dr}', \mathbf{i}_{qr}'$	direct/quadrature component of rotor current
R $_{\rm s}$, L $_{\rm ls}$	stator winding resistance and inductance
R'_{r}, L'_{lr}	rotor winding resistance and inductance
L _m	magnetizing inductance
$\phi_{\rm ds}$, $\phi_{\rm qs}$	direct/quadrature component of stator flux
$\phi_{\rm dr}^{\prime}$, $\phi_{\rm qr}^{\prime}$	direct/quadrature component of rotor flux
L_{s} , L_{r}'	total stator and rotor inductances
T _e	electromagnetic torque
T _L	shaft load torque
ω _m	angular velocity of the rotor
θ _m	rotor angular position
ω,	electrical angular velocity
р	number of pole pairs
J	combined rotor and load inertia coefficient
В	combined rotor and load viscous friction coefficient
R_{c}, L_{c}	cable resistance and inductance

^{*} Prime signs show that all parameters are referred to the stator.

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1. Introduction

Induction Motor (IM), also called asynchronous motor, is the most common motor in various sections of the world including industrial, domestic, educational sections, etc. [1], [2]. Single phase IMs are commonly used in household applications, while three-phase ones are widely used in industry. Threephase IMs are cheap, robust, efficient and reliable; also they have low maintenance cost and high starting torque. In addition, their range is wide (from a few watts to values on the order of 10000 hp) and their speed is nearly constant. In contrast, speed control of IMs is not easy and they have low power factor (lagging) in the lightly loaded condition. Moreover, their starting current may be five to ten times the full-load current [3], [4].

As mentioned above, IMs are very important equipment in the industry. Without induction motors, many products cannot be produced by manufactures and human's life is disrupted. But one of the disadvantages of IMs is the high starting current which can damage stator windings and fluctuate grid voltage. Also, there is need to higher range of switches and equipment. This high starting current is more notable in the high-power IMs. Therefore, this current must be estimated accurately to prevent above problems. Usually, starting current of IMs are estimated based on experimental data from the tests or by calculation in the design stage. In [5], a genetic algorithm is used to estimate IM parameters. Also, parameters of Induction motors are estimated using transient stator current in [6]. Moreover, stator current discharge is used in [7] to estimate parameters of single-phase axial flux induction motors. Also, a non-linear optimization routine-based method is proposed to estimate IM parameters in [8]. In [9], parameters of IM are estimated using current envelope. Also, extended kalman filtering algorithm is proposed to estimate state and parameters of IMs in [10].

This paper presents an intelligent approach to evaluate starting current of induction motors in the design stage or operation situation considering aging effects on the equipment (motor and cable) parameters. Also, this intelligent estimator includes cable effect on the starting current of IMs [5], [6].

In this paper power system blockset (PSB), a MATLAB/Simulink-based simulation tool, is used for calculation of starting current of IMs [7]. In order to study various conditions of starting an IM, many possible system configurations must be considered which needs many time-domain simulations resulting in a large amount of simulation time. This paper presents a real-time estimator for starting current of IMs. The Artificial Neural Network (ANN) is used as intelligent tool

for this purpose. A tool such as proposed in this paper that can give the starting current peak will be helpful to the manufactures during design stage and to the operators during operation condition. The ANN is trained with the most common structures. In the proposed estimator, we have considered the most important aspects which influence the starting current peak such as supply voltage, cable resistance and reactance, stator/rotor resistance and reactance, magnetizing reactance, and switching angle. This information will help the companies and operators to design and handle IMs safely with starting current appearing safe within the limits. Results of the studies shows that developed ANNs can estimate starting current of induction motors with excellent accuracy.

2. Study of System Modelling A. Induction Machine

The electrical section of IM is represented by a fourth-order state-space model and the mechanical section by a second-order system [8]. All electrical variables and parameters are referred to the stator. All stator and rotor quantities are in the arbitrary two-axis reference frame (d-q frame). Equivalent circuit of an induction machine is shown in Fig. (1) Electrical equations for the stator and rotor circuits of the induction machine described in d- and q-axes are as follows.

$$V_{qs} = R_{s}i_{qs} + \frac{d}{dt}\phi_{qs} + \omega\phi_{ds}$$
(1)

$$V_{ds} = R_{s} i_{ds} + \frac{d}{dt} \varphi_{ds} - \omega \varphi_{qs}$$
(2)

$$V'_{qr} = R'_{r}i'_{qr} + \frac{d}{dt}\phi'_{qr} + (\omega - \omega_{r})\phi'_{dr}$$
(3)

$$V'_{dr} = R'_{r}i'_{dr} + \frac{d}{dt}\phi'_{dr} - (\omega - \omega_{r})\phi'_{qr}$$
⁽⁴⁾

$$T_{e} = \frac{3}{2} p (\phi_{ds} i_{qs} - \phi_{qs} i_{ds})$$
(5)

where:

$$\varphi_{qs} = L_s i_{qs} + L_m i'_{qr}$$
(6)

$$\varphi_{ds} = L_{s}i_{ds} + L_{m}i'_{dr}$$
(7)

$$\phi'_{qr} = L'_r i'_{qr} + L_m i'_{qs} \tag{8}$$

$$\varphi'_{\rm dr} = L'_{\rm r} \dot{i}'_{\rm dr} + L_{\rm m} \dot{i}'_{\rm ds} \tag{9}$$

$$\mathbf{L}_{\mathrm{s}} = \mathbf{L}_{\mathrm{ls}} + \mathbf{L}_{\mathrm{m}} \tag{10}$$

$$\mathbf{L}_{\mathrm{r}}^{\prime} = \mathbf{L}_{\mathrm{lr}}^{\prime} + \mathbf{L}_{\mathrm{m}} \tag{11}$$



Fig. (1): Equivalent circuit of induction machine; (a) qaxis, (b) d-axis



current peak

Also, mechanical part equations are:

B. Connecting Cable

Usually cables are modeled using series resistance and inductance and parallel capacitance in the distributed form or in the PI form [9]. As shown in Fig. 2, cable capacitance has negligible effect on the starting current peak. Thus, in this work cable capacitance has been neglected. This capacitor must be used for the long cable or for the transient studies of induction machines where it isn't negligible [10]. Therefore, connecting cable is modeled by a series equivalent resistance and reactance.

3. Study of Stating Current

The electric motive force (EMF) induced in the rotor depends on relative speed of rotor shaft and synchronous speed. Since in the starting condition the motor is standstill, this relative speed has its maximum value and therefore large EMF is induced in the rotor. When an induction motor is stationary, it behaves like a transformer which its secondary winding is short circuited. This causes low impedance to the system voltage and hence IM draws a high current from the gird, typically 5-10 times full-load current.

Motor load doesn't affect starting current peak; although the inertia of the motor and load must be overcome. If inertia is big, the motor takes more time to reach full speed. When the motor accelerates, part of the starting current power overcomes this inertia and is converted to kinetic energy. The remaining power of the starting current heats the rotor, up to possibly 250°C for a "long" starting (20 seconds).

The sample system considered for explanation of the proposed methodology consists a high-power induction motor which produces notable starting current. This system is shown in Fig. 3. For this purpose, induction motor KHV355-2 from VALIADIS company was considered [11]. It is a 200 kW (270 hp), 3300 V induction motor which is in the high-power/medium-voltage category. Parameters of this motor were calculated using noload test, locked-rotor test, and DC test [12]. This motor is fully simulated in the MATLAB software [7]. Fig. 4 shows stator currents of this motor.



Fig. (3): Sample system for the starting current study

In practical system a number of factors affect the starting current peak. In this paper following parameters were considered:

- Supply Voltage (V)
- Equivalent resistance of the connecting cable (R_c)
- Equivalent reactance of the connecting cable (X_c)
- Stator resistance (R_s)
- Stator reactance (X_{ls})
- Rotor resistance (R_r)
- Rotor reactance (X'_{lr})
- Magnetizing reactance (X_m)
- Switching angle (S.A.)

Stator reactance affects the starting current. Fig. 5 shows the effect of stator reactance on the starting current peak at different rotor resistance. Starting current peak is the maximum of stator current peaks between all three phases. Fig. 6 shows the effect of cable resistance on the starting current peak at different supply voltage. Fig. 7 presents effect of stator resistance on the starting current peak at different rotor reactance. Fig. 8 shows the effect of magnetizing reactance on the starting current peak at different stator resistance. Finally, effect of switching angle on the starting current peak has been shown in Fig. 9.



Fig. (5): Starting current peak as stator reactance while supply voltage 1 p.u., cable resistance 0.004 p.u., cable reactance 0.0126 p.u., stator resistance 0.012 p.u., referred rotor reactance 0.1 p.u., magnetizing reactance 2 p.u., and switching angle 0°



Fig. (6): Starting current peak as cable resistance while cable reactance 0.0126 p.u., stator resistance 0.012 p.u., stator reactance 0.08 p.u., referred rotor resistance 0.01 p.u., referred rotor reactance 0.1 p.u., magnetizing reactance 2.2 p.u., and switching angle 0° .







Fig. (7): Starting current peak as stator resistance while supply voltage 0.9 p.u., cable resistance 0.006 p.u., cable reactance 0.0063 p.u., stator reactance 0.12 p.u., referred rotor resistance 0.008 p.u., magnetizing reactance 2.3 p.u., and switching angle 0°



Fig. (8): Starting current peak as magnetizing reactance while supply voltage 1 p.u., cable resistance 0.002 p.u., cable reactance 0.0063 p.u., stator reactance 0.1 p.u., referred rotor resistance 0.012 p.u., referred rotor reactance 0.1 p.u., and switching angle 0°



Fig. (9): Starting current peak as switching angle while supply voltage 1 p.u., cable resistance 0.004 p.u., stator resistance 0.012 p.u., stator reactance 0.1 p.u., referred rotor resistance 0.014 p.u., referred rotor reactance 0.1 p.u., and magnetizing reactance 2 p.u

4. The Artificial Neural Network

There are many types of neural networks for various applications available in the literature [13-17]. Multilayer Perceptrons (MLPs) and Radial Basis Functions (RBFs) are examples of feed-forward networks and both universal approximators. In spite of being different networks in several important respects, these two neural network architectures are capable of accurately mimicking each other.

In this work, different algorithms were used to train MLP structure: Back Propagation (BP), Delta-Bar-Delta (DBD), Extended Delta-Bar-Delta (EDBD), Directed Random. Search (DRS), Quick Propagation (QP), and Levenberg–Marquardt (LM). Because of space limitation, these structures and related algorithms are not described here; detailed structure of BP, DBD, EDBD, DRS, QP, and RBF are presented in [13], and LM algorithm is fully discussed in [18]. The basic structure of developed artificial neural network is shown in Fig. 10.

a) Training of Artificial Neural Network

Parameters listed in Section 3 are adopted as ANN inputs and starting current peak is ANN output. To train ANNs, all experiments have been repeated for different system parameters. For producing learning and testing sets, ANN inputs were varied in different steps (depend on the parameter). 10% of these sets were used for ANN learning and 90% of these sets were used for ANN testing. Each ANN is trained with the goal of mean square error (MSE) 1e-6. Fig. 11 shows the training of developed neural networks. Specifications of ANNs are presented in Table 1. After learning, all parameters of the trained networks have been frozen and then used in the retrieval mode for testing the capabilities of the system on the data not used in learning. The testing data samples have been generated through the PSB program by placing the parameter values not used in

learning, by applying different parameters. A large number of testing data have been used to check the proposed solution in the most objective way at practically all possible parameters variation. Relative error is calculated by the difference of PSB output and ANN output:

$$\operatorname{Er}_{\text{Relative}} (\%) = \frac{\left|I_{\text{ANN}} - I_{\text{PSB}}\right|}{I_{\text{PSB}}} \times 100$$
(14)

and absolute error is calculated as:

$$\mathrm{Er}_{\mathrm{Absolute}} = \left| \mathbf{I}_{\mathrm{ANN}} - \mathbf{I}_{\mathrm{PSB}} \right| \tag{15}$$

where I_{ANN} is starting current peak calculated by ANN, and I_{PSB} refers to starting current peak calculated by PSB.



Fig. (10): Basic structure of developed artificial neural network

Table (1): Specifications of developed ANNs

ANN model	Number of neurons in the hidden layer	Training time [epochs]				
BP	5	178				
DBD	7	1189				
EDBD	8	315				
DRS	6	109				
LM	6	286				
QP	7	975				
RBF	5	354				



Fig. (11): Squared error against epoch curve for developed ANNs

Fig. 12 shows relative errors for LM algorithm. Moreover, results for a sample test data for all developed ANNs are presented in Table 2 and Figs. 13-15. Calculated errors for different ANNs in Table 2 are relative errors. Fig. 13 shows starting current peak against the supply voltage, Fig. 14 presents starting current peak against the cable reactance, and Fig. 15 shows starting current peak against the stator resistance.

5. Discussion

In this paper, starting current peak was evaluated using BP, DBD, EDBD, DRS, QP, LM and RBF neural networks. To select best approach for starting current evaluation, a comparison has been made. Table 3 presents a comparison between these methods based on average of relative and absolute errors for Table 2 sample data. It can be seen from Table 3 that LM and EDBD algorithms have better performance (smaller relative and absolute errors) to evaluate starting current peak in the induction motors.

6. Conclusion

This paper presents an artificial neural networkbased approach to evaluate starting current peak of induction motors including cable effect. Both MLP and RBF structures have been employed for this

purpose. MLP is trained with BP, DBD, EDBD, DRS, QP, and LM algorithms. Simulation results show that most developed ANNs can estimate starting current; however LM and EDBD algorithms present better accuracy. This technique can help the companies and operators to evaluate starting current peak during both design and operation stages in real-time.

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Table (3): Average of Relative and Absolute Errors f	for
Table (2) Sample Data	

ANN Model	Average of Relative Error [%]	Average of Absolute Error [A]
LM	0.0922	0.4034
EDBD	0.1272	0.5797
BP	0.1824	0.7716
DBD	0.2114	0.9137
QP	0.4066	1.7282
RBF	0.7624	3.3372
DRS	1.9902	8.9127



Fig. (12): Relative errors of learning and testing sets for LM algorithm

v	0.8	0.8	0.8	0.8	0.9	0.9	0.9	0.9	1	1	1	1	1.1	1.1	1.1	1.1
R _c	0.6	0.6	0.6	0.6	0.2	0.2	0.2	0.2	0.3	0.4	0.5	0.6	0.3	0.3	0.3	0.3
X _c	0.031 4	0.031 4	0.031 4	0.031 4	0.018 9	0.025	0.031 4	0.037 7	0.025	0.025	0.025	0.025	0.031 4	0.031 4	0.031 4	0.031
R _s	0.5	0.4	0.4	0.5	0.6	0.6	0.6	0.6	0.5	0.5	0.5	0.5	0.4	0.5	0.6	0.7
X _{ls}	4	4	4	4	3	4	5	6	5	5	5	5	4	4	4	4
R' _r	0.4	0.4	0.4	0.4	0.4	0.6	0.6	0.4	0.4	0.5	0.6	0.7	0.6	0.6	0.6	0.6
X_{lr}^{\prime}	3	4	5	6	5	5	5	5	5	5	5	5	3	3	3	3
Xm	110	110	110	110	90	100	110	120	100	100	100	100	110	110	110	110
S.A.[°]	10	10	10	10	30	30	30	30	55	55	55	55	80	80	80	80
PSB	409.3 465	381.9 27	354.2 365	327.2 188	459.0 562	398.8 71	368.4 783	344.4 66	420.7 591	411.5 882	403.1 649	395.2 277	581.0 046	573.7 519	566.5 372	559.3 91
LM	409.5 152	382.5 548	354.1 394	326.7 324	458.6 114	398.1 558	368.4 264	344.7 404	420.8 398	412.3 009	403.1 615	394.6 765	580.1 499	572.8 548	566.4 511	559.7 936
Error	0.041 216	0.164 401	0.027 428	0.148 647	0.096 902	0.179 304	0.014 072	0.079 682	0.019 197	0.173 142	0.000 834	0.139 484	0.147 115	0.156 365	0.015 198	0.071 961
EDB D	408.9 137	382.0 169	354.7 666	327.6 04	458.2 575	399.5 689	369.0 665	344.4 949	420.8 314	411.2 594	403.6 999	395.8 744	581.5 967	572.5 757	567.5 547	560.7 457
Error	0.105 721	0.023	0.149 631	0.117 731	0.173 987	0.174 972	0.159 633	0.008 401	0.017 202	0.079 9	0.132 716	0.163 611	0.101 905	0.204 995	0.179 59	0.242 162
BP	408.3	382.1 378	355.2 442	326.1 499	458.2 769	397.7 07	369.1 933	344.6 419	420.4	411.0	402.1	396.3 045	579.4 9	573.1 488	565.5 388	559.5 571
Error	0.253	0.055	0.284	0.326	0.169	0.291	0.194 049	0.051	0.065	0.134 295	0.247	0.272	0.260	0.105	0.176	0.029
DBD	409.0 761	383.1 141	352.9 714	327.6 894	460.3	399.3 52	369.2 799	343.2 903	419.7	412.1 541	402.6 703	395.9 611	579.9 978	574.5 978	567.8 341	561.0 941
Error	0.066	0.310 826	0.357	0.143 818	0.281	0.120 601	0.217	0.341 294	0.244 971	0.137 478	0.122 682	0.185	0.173 285	0.147 439	0.228 911	0.304 444
QP	411.2 067	379.5 639	351.7 518	325.8 739	462.3 51	400.4 607	367.1 107	345.2 861	422.0 617	410.1 308	402.9 587	392.7 016	581.2 75	575.5 538	569.9 091	557.8 02
Error	0.454 422	0.618 711	0.701 44	0.411 004	0.717 732	0.398 55	0.371 13	0.238 091	0.309 601	0.354 101	0.051 147	0.639 172	0.046 536	0.314 053	0.595 173	0.284 065
RBF	407.6	380.9	358.8 303	326.9 064	462.7 908	395.3 13	371.8	348.6	421.0 887	416.2	400.1	399.0 697	579.2 769	569.3 88	572.1 093	551.9 008
Error	0.412	0.251	1.296 812	0.095	0.813	0.892	0.911	1.210	0.078	1.143	0.740	0.972	0.297	0.760	0.983	1.339
DRS	412.7	373.4	343.9	328.1	474.2	399.2 076	378.9	354.2	420.0	403.9	416.0	399.5 476	596.9 357	585.4 837	551.3 274	574.5
Error	0.822	2.211 562	2.915 072	0.286 86	3.305 103	0.106 964	2.840 375	2.841 425	0.169 516	1.856 013	3.206 777	1.093 008	2.741 987	2.044 756	2.684 71	2.717 229

Table (2): Some Sample Testing Data and Output



Fig. (13): Starting current peak vs. supply voltage simulated by ANNs and PSB while cable resistance 0.006 p.u., cable reactance 0.0188 p.u., stator resistance 0.012 p.u., stator reactance 0.06 p.u., referred rotor resistance 0.012 p.u., referred rotor reactance 0.1 p.u., magnetizing reactance 2.2 p.u., and switching angle 0°



Fig. (14): Starting current peak vs. cable reactance simulated by ANNs and PSB while supply voltage 1 p.u., cable resistance 0.004 p.u., stator resistance 0.01 p.u., stator reactance 0.08 p.u., referred rotor resistance 0.008 p.u., referred rotor reactance 0.1 p.u., magnetizing reactance 2.4 p.u., and switching angle 0°



Fig. (15): Starting current peak vs. stator resistance simulated by ANNs and PSB while supply voltage 0.9 p.u., cable resistance 0.002 p.u., cable reactance 0.0188 p.u., stator reactance 0.1 p.u., referred rotor resistance 0.012 p.u., referred rotor reactance 0.1 p.u., referred rotor reactance 0.012 p.u.,

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