# Rotated Unscented Kalman Filter for Two State Nonlinear Systems

Mohammad Esmaeil Akbari <sup>1</sup>, Sepideh Tashahodjamal <sup>2</sup>

<sup>1</sup>Department of Electrical Engineering, Ahar Branch, Islamic Azad University, Ahar, Iran Email: M-Akbari@iau-Ahar.ac.ir (Corresponding author)

Department of Electrical Engineering, Ahar Branch, Islamic Azad University, Ahar, Iran Email: S-Tashahodjamal@iau-Ahar.ac.ir

### **ABSTRACT**

In the several past years, Extended Kalman Filter (EKF) and Unscented Kalman Filter (UKF) have became basic algorithm for state-variables and parameters estimation of discrete nonlinear systems. The UKF has consistently outperformed for estimation. Sometimes least estimation error doesn't yield with UKF for the most nonlinear systems. In this paper, we use a new approach for a two variable state nonlinear systems which it is called Rotated UKF (R\_UKF). R\_UKF can be reduced estimation error and reached for least error in state estimation.

KEYWORDS: Extended kalman filter (EKF), unscented kalman filter (UKF), rotated UKF (R-UKF)

### 1. INTRODUCTION

It is derived from the Kalman filter based on successive linearization of the signal process and observation map [1]. The algorithm of EKF is sub-optimal, so can be became divergence. UKF is a freedivergence approach and it can be used instead of EKF for state estimation [2]. The UKF has been used for state-estimation and parameter-estimation at nonlinear systems. The UKF is a derivative free alternative to the EKF. R UKF is a new suggestion algorithm for estimation two variable state nonlinear systems. This paper is organized as follows. After the introduction in section 1, discussion UKF algorithm is discussion for estimation of parameter nonlinear system in section 2, section 3 present equal R UKF for nonlinear system with two state variables. Next approach R UKF and UKF

for as sample system and compare simulation result in section 4. Finally the conclusions is given in section5.

### 2. UNSCENTED KALMAN FILTER

The basic usage for the UKF is estimation of the state of a discrete-time nonlinear dynamic system. The UKF utilizes the Unscented Transformation (UT). For this application, the system nonlinear state equations are expressed in the discretized form:

$$X_{K+1} = F(X_K, U_K) + V_K$$
  

$$Y_K = H(X_K) + n_K$$
 (1)

 $X_K$  as a parameter that represents the state-variable of the system,  $U_K$  is a known input and  $Y_K$  is the output measurement signal. The process noise  $V_K$  drives dynamic system and the measurement noise is given by  $N_K$ .  $V_K$  and  $N_K$  in the UKF are

assumed Gaussian noise. Sometimes UKF is used for system identification, with nonlinear mapping  $Y_K = G(X_K, W)$ , where  $X_K$  is input,  $Y_K$  is output, and the nonlinear map G(.) is parameterized by vector W.

$$W_{K+1} = W_K + r_K$$
  
 $d_K = G(X_K, W_K) + e_K$  (2)

The process noise  $r_{K}$  drives the dynamic system, and the measurement noise is given by  $e_K$ . The output  $d_K$  corresponds to a nonlinear observation on  $W_{\kappa}$ . The EKF have inherent flaws, because EKF uses of linearization approach for calculating the mean and covariance of a random variables [2,3,4]. In the UKF eliminated these flaws by using a deterministic sampling approach covariance calculate and mean. Essentially, 2L+1 sigma points are chosen based on a square-root decomposition of the previous covariance. These sigma points are spread to the true nonlinearity, without approximation, and then a weighted mean and covariance is calculated. A simple illustration of this approach is shown in Figure 1 for a two-dimensional system. Figure 1-a shows the true mean and covariance propagation using Monte-Carlo sampling; the center plots show the results using a linearization approach (EKF) , the right plots show the performance of the UKF approach for 5 sigma points. The standard UKF involves the recursive application of this sampling approach to the state-space equations. The standard UKF shown in Algorithm 1 for state-estimation, with using the following definitions:

$$W_0^m = \lambda / (L + \lambda)$$

$$W_0^c = \lambda / (L + \lambda) + (1 - \alpha^2 + \beta)$$

 $W_i^c = W_i^m = 1/\{2(L+\lambda)\}$  i = 1, 2, ..., 2L (3)

 $W_i$  Is the weight associated with the i-th sigma point so that  $\sum_{i=1}^{2L} W_i = 1$ .

Scaling parameters are:

$$\lambda = \alpha^{2} (L + \kappa) - L$$
$$\gamma = \sqrt{(L + \lambda)}$$

$$\kappa = 3 - L$$
 (4)

The constant  $\alpha$  determines the spread of the sigma points around  $\hat{X}(10^{-4} \le \alpha \le 1)$ .  $\kappa$  is second scaling parameter which is usually set to  $\kappa = 3 - L$  and  $\beta$  is used to incorporate prior knowledge of the distribution of X (for Gaussian distributions),  $\beta = 2$  is optimal Also note that we use linear algebra operation by adding a column to a matrix. Now it is choosed a set of 2L+1 weighted samples  $X_i$  (sigma points) deterministically so that they completely represent the true mean and covariance of state X.

$$\begin{split} \hat{X}_{0} &= \hat{X} \\ \hat{X}_{i} &= \hat{X} + (\gamma \sqrt{P_{K-1}})_{i} & i = 1, ..., L \\ \hat{X}_{i} &= \hat{X} - (\gamma \sqrt{P_{K-1}})_{i} & i = n+1, ..., L \\ X_{k-1} &= \begin{bmatrix} \hat{X}_{k-1} & \hat{X}_{k-1} + (\gamma \sqrt{P_{k-1}}) & \hat{X}_{k-1} - (\gamma \sqrt{P_{k-1}}) \end{bmatrix} \end{split}$$
(5)

The mean and covariance of *Y* are approximated by the weighted average mean and covariance of the transformed sigma point.

$$\hat{y} = \sum_{i=0}^{2L} W_i Y_i$$

$$P_y = \sum_{i=0}^{2L} W_i (Y_i - \hat{y}) (Y_i - \hat{y})^T$$
(6)

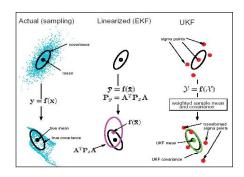
Figure 2 shows the standard UKF algorithm.

# 3. ROTATED UNSCENTED KALMAN FILTER

However, standard UKF approach has more advantages with comparing EKF [3,4], but it couldn't calculate state variables without error and bios. In this paper we discuss a new algorithm for calculating sigma points, which it chooses sigma points better than UKF. Percentage of improving is correlated to type of nonlinear system [5]. This new algorithm can be used for nonlinear systems with only two state variables.

$$X_{K+1} = F(X_K, U_K) + V_K$$
  

$$Y_K = H(X_K) + n_K$$
(7)



**Fig.1.** Example of mean and covariance propagation.

Initial with:

 $\hat{X}_{0} = E[X_{0}] \qquad P_{0} = E[(X_{0} - \hat{X}_{0})(X_{0} - \hat{X}_{0})^{T}]$ For  $K \in \{1, 2, ...., \infty\}$ Calculate Sigma points:

1:  $X_{k-1} = [\hat{X}_{k-1} \quad \hat{X}_{k-1} + (y\sqrt{P_{k-1}}) \quad \hat{X}_{k-1} - (y\sqrt{P_{k-1}})]$ Time update:

2:  $X_{k|k-1}^{*} = F[X_{k-1}, U_{k-1}]$ 3:  $\hat{X}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{m} X_{i,k|k-1}^{*}$ 4:  $P_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{c} [X_{i,k|k-1}^{*} - \hat{X}_{k}^{-}][X_{i,k|k-1}^{*} - \hat{X}_{k}^{-}]^{T} + R^{v}$   $R^{v} = \text{process noise covariance}$ 5:  $X_{k|k-1} = [\hat{X}_{k}^{-} \quad \hat{X}_{k}^{-} + (y\sqrt{P_{k}^{-}}) \quad \hat{X}_{k}^{-} + (y\sqrt{P_{k}^{-}})]$ 6:  $Y_{k|k-1} = H[X_{k|k-1}]$ 7:  $\hat{Y}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{m} Y_{i,k|k-1}$ Measurement updates equations:

8: 
$$P_{\hat{y}_k\hat{y}_k} = \sum_{i=0}^{2L} W_i^c \left[ Y_{i,k|k-1} - \hat{y}_k^- \right] \left[ Y_{i,k|k-1} - \hat{y}_k^- \right]^T + R^n$$
 $R^n =$  measurement noise covariance

9:  $P_{x_ky_k} = \sum_{i=0}^{2L} W_i^c \left[ X_{i,k|k-1} - \hat{x}_k^- \right] \left[ Y_{i,k|k-1} - \hat{y}_k^- \right]^T$ 

10:  $K_k = P_{x_ky_k} P_{\hat{y}_k\hat{y}_k}^{-1}$ 

11:  $\hat{X}_k = \hat{X}_k^- + K_k \left( y_k - \hat{y}_k^- \right)$ 

12:  $P_k = P_k^- - K_k P_{\hat{y}_k\hat{y}_k} K_k^T$ 

Fig.2. Standard UKF algorithm.

State variables: 
$$X = \begin{bmatrix} X_1 \\ X_2 \end{bmatrix}$$

For two state variables nonlinear systems the UKF algorithm calculates sigma points with equations (8):

$$\hat{X}_0 = E[X_0] \qquad P_0 = E[(X_0 - \hat{X}_0)(X_0 - \hat{X}_0)]$$

$$\gamma \sqrt{P_{k-1}} = \begin{bmatrix} a1 & a2\\b1 & b2 \end{bmatrix}$$
(8)

 $P_{K-1}$  is state covariance matrix:

$$\begin{pmatrix} \gamma \sqrt{P_{k-1}} \end{pmatrix}_{1} = \begin{bmatrix} a1 \\ b1 \end{bmatrix} \\
\begin{pmatrix} \gamma \sqrt{P_{k-1}} \end{pmatrix}_{2} = \begin{bmatrix} a2 \\ b2 \end{bmatrix}$$
(9)

Sigma points are follows:

$$X_{k-1} = \left[ \hat{X}_{k-1} \ \hat{X}_{k-1} + \left( y \sqrt{P_{k-1}} \right)_{1} \ \hat{X}_{k-1} + \left( y \sqrt{P_{k-1}} \right)_{2} \ \hat{X}_{k-1} - \left( y \sqrt{P_{k-1}} \right)_{1} \ \hat{X}_{k-1} - \left( y \sqrt{P_{k-1}} \right)_{2} \right] \ (10)$$

It is used equation (11) for calculating update sigma points:

$$X_{k|k-1} = \begin{bmatrix} \hat{X}_k & \hat{X}_k^- + \left( \sqrt{P_k} \right) & \hat{X}_k^- + \left( \sqrt{\sqrt{P_k^-}} \right) & \hat{X}_k^- - \left( \sqrt{P_k^-} \right) & \hat{X}_k^- - \left( \sqrt{P_k^-} \right) \end{bmatrix} \uparrow (11)$$

In  $R\_UKF$  , it is utilized following equations for sigma points calculating:

Let:

$$M\theta 1 = \begin{bmatrix} \cos(\theta) \\ \sin(\theta) \end{bmatrix} & \& \quad M\theta 2 = \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$
 (12)

In these equations  $\theta$  can be represented by:  $0 \le \theta \le 180^{\circ} \text{ Or } -90^{\circ} \le \theta \le +90^{\circ}$ 

New sigma points are follows:

$$X_{k-1} = \begin{bmatrix} \hat{X}_{k-1} \\ \hat{X}_{k-1} + (\gamma \sqrt{P_{k-1}})_{1} . M \theta 1 \\ \hat{X}_{k-1} + (\gamma \sqrt{P_{k-1}})_{2} . M \theta 2 \\ \hat{X}_{k-1} - (\gamma \sqrt{P_{k-1}})_{1} . M \theta 1 \\ \hat{X}_{k-1} - (\gamma \sqrt{P_{k-1}})_{2} . M \theta 2 \end{bmatrix}$$
(13)

Also, equation (14) in indicates to update sigma points:

$$X_{k-1} = \begin{bmatrix} \hat{X}_{k-1} + (\gamma \sqrt{P_k^-})_1 . M\theta 1 \\ \hat{X}_{k-1} + (\gamma \sqrt{P_k^-})_2 . M\theta 2 \\ \hat{X}_{k-1} - (\gamma \sqrt{P_k^-})_1 . M\theta 1 \\ \hat{X}_{k-1} - (\gamma \sqrt{P_k^-})_2 . M\theta 2 \end{bmatrix}$$
(14)

## R UKF approach is shown in figure 2.

Initial with:  

$$X = \begin{bmatrix} X_1 \\ X_2 \end{bmatrix} \hat{X}_0 = E[X_0] P_0 = E[(X_0 - \hat{X}_0)(X_0 - \hat{X}_0)]$$
For  $K \in \{1, 2, ..., \infty\}$   

$$M\theta 1 = \begin{bmatrix} \cos(\theta) \\ \sin(\theta) \end{bmatrix} & M\theta 2 = \begin{bmatrix} -\sin(\theta) \\ \cos(\theta) \end{bmatrix}$$
Calculate Sigma points:

1: 
$$X_{k-1} = \begin{bmatrix} \hat{X}_{k-1} \\ \hat{X}_{k-1} + (\gamma \sqrt{P_{k-1}})_1 . M \theta 1 \\ \hat{X}_{k-1} + (\gamma \sqrt{P_{k-1}})_2 . M \theta 2 \\ \hat{X}_{k-1} - (\gamma \sqrt{P_{k-1}})_1 . M \theta 1 \\ \hat{X}_{k-1} - (\gamma \sqrt{P_{k-1}})_2 . M \theta 2 \end{bmatrix}$$

Time update:

2: 
$$X_{k|k-1}^* = F[X_{k-1}, U_{k-1}]$$
  
3:  $\hat{X}_k^- = \sum_{i=0}^{2L} W_i^m X_{i,k|k-1}^*$ 

4: 
$$P_k^- = \sum_{i=0}^{2L} W_i^c \left[ X_{i,k|k-1}^* - \hat{X}_k^- \right] \left[ X_{i,k|k-1}^* - \hat{X}_k^- \right]^T + R^v$$

 $R^{v}$  = process noise covariance

5: 
$$X_{k-1} = \begin{bmatrix} \hat{X}_{k-1} \\ \hat{X}_{k-1} + (\gamma \sqrt{P_k^-})_1 . M\theta 1 \\ \hat{X}_{k-1} + (\gamma \sqrt{P_k^-})_2 . M\theta 2 \\ \hat{X}_{k-1} - (\gamma \sqrt{P_k^-})_1 . M\theta 1 \\ \hat{X}_{k-1} - (\gamma \sqrt{P_k^-})_2 . M\theta 2 \end{bmatrix}$$
6: 
$$Y_{k|k-1} = H[X_{k|k-1}]$$
7: 
$$\hat{Y}_k^- = \sum_{i=0}^{2L} W_i^m Y_{i,k|k-1}$$

Measurement updates equations:

8: 
$$P_{\hat{y}_k \hat{y}_k} = \sum_{i=0}^{2L} W_i^c \left[ Y_{i,k|k-1} - \hat{y}_k^- \right] \left[ Y_{i,k|k-1} - \hat{y}_k^- \right]^T + R^n$$

 $R^n$  = measurement noise covariance

9: 
$$P_{x_k y_k} = \sum_{i=0}^{2L} W_i^c \left[ X_{i,k|k-1} - \hat{x}_k^- \right] \left[ Y_{i,k|k-1} - \hat{y}_k^- \right]^T$$

10: 
$$K_k = P_{x_k y_k} P_{\hat{y}_k \hat{y}_k}^{-1}$$

11: 
$$\hat{X}_k = \hat{X}_k^- + K_k (y_k - \hat{y}_k^-)$$

12: 
$$P_k = P_k^- - K_k P_{\hat{y}_k \hat{y}_k} K_k^T$$

Fig.3. Rotate UKF algorithm.

For more nonlinear systems can find convenient rotate angle, which errors are minimum [6]. We called  $\theta$  this rotation angle.  $\theta$  can add to adjusting parameters of standard UKF algorithm [7].

### 4. SIMULATION

The improvement in error performance of the R\_UKF for state estimation is shown in following example. Results of simulation for R\_UKF are compared with standard UKF.

$$\dot{X}_{0} = -X_{0} - 2X_{1}^{2} + U$$

$$\dot{X}_{1} = X_{0}X_{1} - X_{1}^{3} + U$$

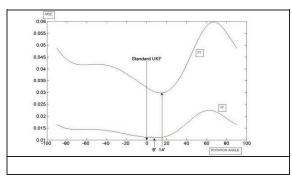
$$Y = 0.8X_{0} + X_{1}$$
Initial conditions are:
$$P_{0} = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix} \quad X_{0} = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

$$R_{0} = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix} \quad R^{n} = 0.12$$

(16)

Figure 4 shows relation between Mean Square Error (MSE) of  $X_0$  &  $X_1$  and rotated angle. For this example, least MSE for state  $X_0$  is at rotated angle  $14^{\circ}$  and least MSE for state  $X_1$  is at rotated  $8^{\circ}$ . Standard UKF is equal with rotated  $0^{\circ}$ . In this nonlinear system, the best rotated angle is about  $11^{\circ}$  for total least MSE.

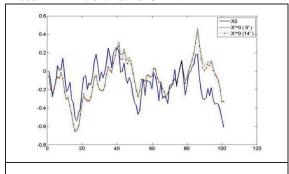
 $\alpha = 1$   $\beta = 2$   $\kappa = 0$ 



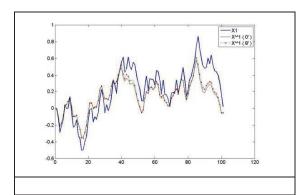
**Fig.4.** MSE of  $X_0$  &  $X_1$  due rotation angle from  $-90^{\circ}$  to  $+90^{\circ}$ 

Figure 5 & Figure 6 show the superior performance of R\_UKF compared with standard UKF for state estimating the white noise (3dB SNR). The performance of R\_UKF is superior to the standard UKF for

state estimating. The computational of R-UKF is a few complexes, but MSE for states in the R\_UKF is approximately 10% less than the standard UKF.



**Fig.5.** Estimation of  $X_0$  with white noise (3dB SNR) with standard UKF (at  $0^{\circ}$ ) & R UKF (at  $14^{\circ}$ ).



**Fig.6.** Estimation of  $X_1$  with white noise (3dB SNR) with standard UKF (at  $0^{\circ}$ ) & R\_UKF (at  $8^{\circ}$ ).

### 5. CONCLUSIONS

The R\_UKF consistently performs better than or equal than the standard UKF, with the added advantage of R\_UKF to standard UKF we can perform best estimation for nonlinear systems. In this paper, it is introduced rotated forms of the UKF. If rotation angle of each nonlinear system

adjust finely, the R\_UKF has better properties. Nonetheless, this paper discusses for 2 state nonlinear systems, but it may expand to higher number of state nonlinear systems.

### REFERENCES

- [1] Simon Haykin, Kalman Filtering and neural networks. Communications Research Laboratory, McMaster University, Hamilton, Ontario, Canada, John Wiley & Sons, Inc. NewYork, ISBN 0-471-36998-5.
- [2] S. J. Julier and J. K. Uhlmann, "A New Extension of the Kalman Filter to Nonlinear Systems," in Proc. of Aero Sense: The 11th Int. Symp. On Aerospace/Defense Sensing, Simulation and Controls., 1997.
- [3] E.Wan, R. van derMerwe, and A. T. Nelson, "Dual Estimation and the Unscented Transformation," in Neural Information Processing Systems 12. 2000, pp. 666–672, MIT Press.
- [4] E. A. Wan and R. van der Merwe, "The Unscented Kalman Filter for Nonlinear Estimation," in Proc. of IEEE Symposium 2000 (AS-SPCC), Lake Louise, Alberta, Canada, Oct. 2000.
- [5] S.Gannot, D. Burshtein, and E. Weinstein, "Iterative and Sequential Kalman Filter-Based Speech enhancement Algorithms," IEEE Trans. on Speech and Audio Proc., vol. 6, no. 4, pp. 373–385, Jul. 1998.
- [6] J. L. Crassidis and J. L. Junkins," Optimal Estimation of Dynamic Systems". Boca Raton, Florida: CRC Press, to be published 2004.
- [7] S. J. Julier, "The Scaled Unscented Transformation," in Proceedings of the American Control Conference, vol. 6, pp. 4555–4559, 2002.